A Framework for Asynchronous Circuit Modeling and Verification in ACL2

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Outline

- Introduction
- 2 The DE System
- Modeling and Verification Approach
- 4 Case Studies
- 5 Future Work and Conclusions

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Synchronous circuits (or clock-driven circuits): changes in the state of storage elements are synchronized by **a global clock signal**.

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Why asynchronous?

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Why asynchronous?

- Low power consumption,
- High operating speed,
- Elimination of clock skew problems,
- Better composability and modularity for large systems,
- ...

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Our goal: developing scalable methods for reasoning about the functional correctness of self-timed systems using ACL2.

- We use the DE system [Hunt:2000], which is built in ACL2, to specify and verify self-timed circuit designs.
- Developing a hierarchical verification approach to support scalability.
- Exploring strategies for reasoning with non-deterministic circuit behavior.

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- These lemmas are used to prove the correctness of yet larger modules containing these submodules, without the need to dig into any details about the submodules.
- This approach has been demonstrated its **scalability** to large systems, as shown on contemporary x86 designs at Centaur Technology [Slobodova et al.:2011].

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 - ⇒ Employing an oracle, which we call a collection of go signals

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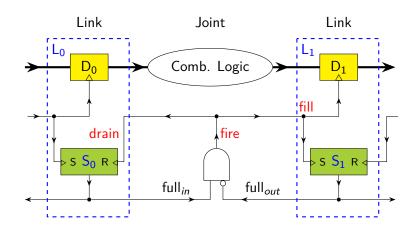
- Links are communication channels in which data and full/empty states are stored.
- Joints are handshake components that implement flow control and data operations.

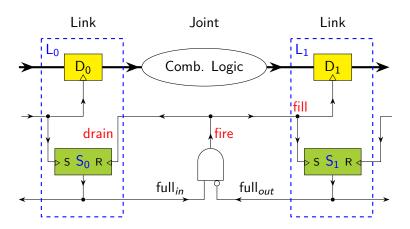
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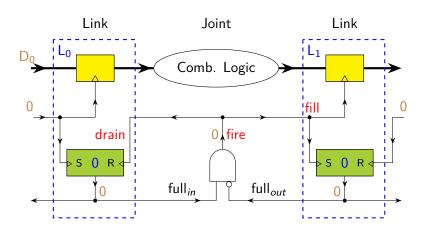
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Joints are the meeting points for links to **coordinate states** and **exchange data**.

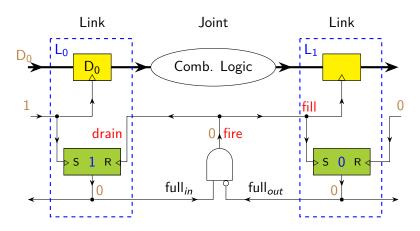




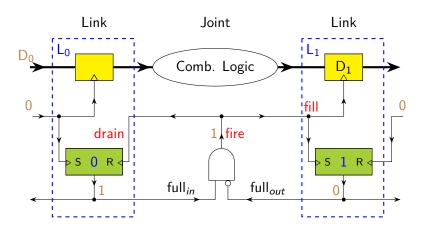
A joint can have several input and output links connected to it.



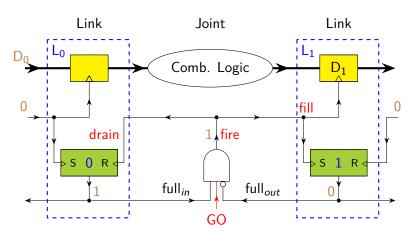
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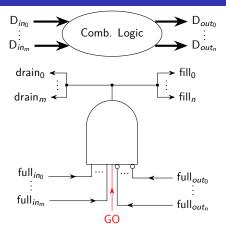
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When a joint fires, the following three actions will be executed in parallel:

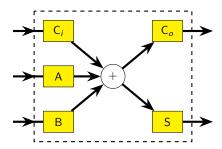
- transfer data computed from the input links to the output links,
- fill the output links, make them full,
- drain the input links, make them empty.

Verification

Our framework applies a hierarchical verification approach to formalizing single transitions of circuit behavior (simulated by se and de functions).

- The output and next state of a module are formalized using the formalized outputs and next states of submodules, without delving into details about the submodules.
- Self-timed modules can be abstracted as "complex" links or "complex" joints.

Self-Timed Modules



A complex link: an adder



A complex joint: a queue Q2 of two links

Verification

Compositional reasoning:

- Functional properties of self-timed systems may involve multi-step executions that are quite burdensome to establish directly.
- Decompose the executions into smaller steps in such a way that sub-properties after executing each of these smaller steps can be carried out much easier.
- The desired properties are then established by simply composing these sub-properties.

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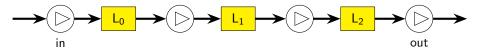
Induction:

- We use induction to prove properties of systems over time, for instance, the relationship between input and output sequences.
- We also apply induction to establishing loop invariants of iterative circuits, i.e., circuits with feedback loops.

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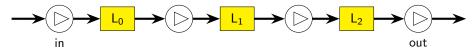
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Q3



For self-timed circuits with no feedback loops, we verify their functional correctness in terms of **the relationship between their input and output sequences**.

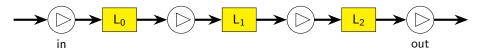
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For self-timed circuits with no feedback loops, we verify their functional correctness in terms of **the relationship between their input and output sequences**.

Let **in-act** denote the **fire** signal from the AND gate in the control logic of joint **in**, **out-act** denote the **fire** signal from the AND gate in the control logic of joint **out**.

Q3

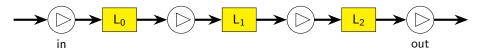


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Let **in-act** denote the **fire** signal from the AND gate in the control logic of joint **in**, **out-act** denote the **fire** signal from the AND gate in the control logic of joint **out**.

Module Q3 will accept a new input if the **in-act** signal is high. In this case, we call this input valid. We define a (valid) input sequence as a sequence of valid inputs.

Q3



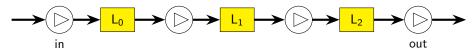
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Let **in-act** denote the **fire** signal from the AND gate in the control logic of joint **in**, **out-act** denote the **fire** signal from the AND gate in the control logic of joint **out**.

Module Q3 will accept a new input if the **in-act** signal is high. In this case, we call this input valid. We define a (valid) input sequence as a sequence of valid inputs.

Similarly, we define a (valid) output sequence as a sequence of valid outputs. Q3 will report a valid output when the **out-act** signal is high.

Q3



We define the function q3\$extract-data(st) that extracts valid data from state st of Q3, i.e. extracts data from links that are **full** at state st. The following equality states the functional correctness of Q3:

```
q3\$extract-data(q3\$run(input-list, st, n)) ++ out-seq = in-seq ++ q3\$extract-data(st)
```

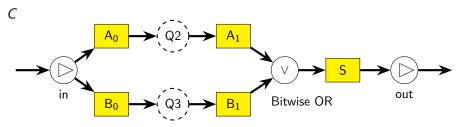
$$q3$$
\$extract-data($q3$ \$run(input-list, st, n)) ++ out-seq =
 in-seq ++ $q3$ \$extract-data(st) (1)

Our ACL2 proof of (1) uses **induction** and the following single-step-update property of Q3 as a supporting lemma:

$$q3\$extract-data(q3\$step(input, st)) = q3\$step-spec(input, st)$$
 (2)

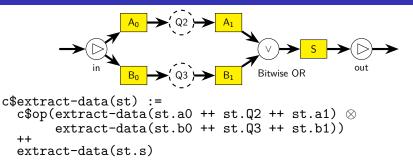
where q3\$step-spec(input, st) :=

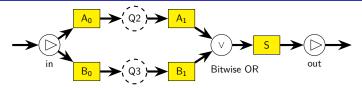
$$\begin{cases} q3\$ extract\text{-}data(st), & \text{if } in\text{-}act = F \land out\text{-}act = F \\ [input] ++ & q3\$ extract\text{-}data(st), & \text{if } in\text{-}act = T \land out\text{-}act = F \\ remove\text{-}last(q3\$ extract\text{-}data(st)), & \text{if } in\text{-}act = F \land out\text{-}act = T \\ [input] ++ & remove\text{-}last(q3\$ extract\text{-}data(st)), & \text{otherwise} \end{cases}$$



In terms of input-output relationship, C simply performs the bitwise OR operation on the two input operands. The operation of C over the input sequence is defined as follows:

```
c$op(in-seq) :=
  if (in-seq is empty) nil
  else {
    let {
      input := first(in-seq)
      a := first(input)
      b := second(input)}
    [v-or(a, b)] ++ c$op(rest(in-seq))}
```





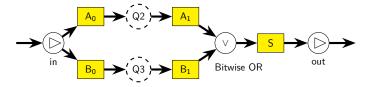
c\$extract-data(st) :=
$$c$op(extract-data(st.a0 ++ st.Q2 ++ st.a1) \otimes extract-data(st.b0 ++ st.Q3 ++ st.b1))$$
 ++ $extract-data(st.s)$

If the current state *st* of *C* satisfies the following condition:

$$size(extract-data(st.a0 ++ st.Q2 ++ st.a1)) = size(extract-data(st.b0 ++ st.Q3 ++ st.b1))$$
(3)

then the following functional property of C holds:

$$c\$extract-data(c\$run(input-list, st, n)) ++ out-seq = \\ c\$op(in-seq) ++ c\$extract-data(st)$$
 (4)

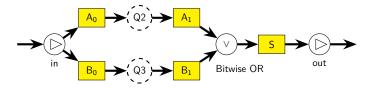


Our ACL2 proof of (4) uses **induction** and the single-step-update property (described below, given that (3) holds) as a supporting lemma.

$$c$$
\$extract-data(c \$step(input, st)) = c \$step-spec(input, st) (5)

where c\$step-spec(input, st) :=

```
\begin{cases} c\$extract\text{-}data(st), in\text{-}act = F \land out\text{-}act = F \\ [v\text{-}or(input.a, input.b)] ++ c\$extract\text{-}data(st), in\text{-}act = T \land out\text{-}act = F \\ remove\text{-}last(c\$extract\text{-}data(st)), in\text{-}act = F \land out\text{-}act = T \\ [v\text{-}or(input.a, input.b)] ++ remove\text{-}last(c\$extract\text{-}data(st)), otherwise \end{cases}
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Our ACL2 proof of (4) uses **induction** and the single-step-update property (described below, given that (3) holds) as a supporting lemma.

$$c\$extract-data(c\$step(input,st)) = c\$step-spec(input,st)$$
 (5)

Our proof of (5) does not concern the details of Q2 and Q3. All we need to know about these two modules in proving (5) is their single-step-update properties.

In other words, we employ a hierarchical strategy in proving the single-step-update property of a self-timed module.

In summary, for each self-timed module that has no feedback loops, we prove the following two properties:

- the single-step-update property (proved by using hierarchical reasoning),
- ② the relationship between the input and output sequences (proved by using induction and the single-step-update property).

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 Computing loop invariants in these systems becomes much more complicated than in synchronous systems.

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We impose design restrictions on iterative circuits to reduce non-determinism, and consequently reduce the complexity of the set of execution paths:

 These restrictions enable our framework to verify loop invariants efficiently via induction and subsequently verify the functional correctness of self-timed circuit designs.

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 These restrictions enable our framework to verify loop invariants efficiently via induction and subsequently verify the functional correctness of self-timed circuit designs.

Design restrictions: A module is ready to communicate with other modules only when it finishes all of its internal operations and becomes quiescent.

32-Bit Self-Timed Serial Adder Verification

We demonstrate our framework by modeling and verifying the functional correctness of a 32-bit self-timed serial adder [Chau:2017].

We prove that the self-timed serial adder indeed performs the addition under an appropriate initial condition.

 When the adder finishes its execution, the result is proven to be the sum of the two 32-bit input operands and the carry-in.

32-Bit Self-Timed Serial Adder Verification

We demonstrate our framework by modeling and verifying the functional correctness of a 32-bit self-timed serial adder [Chau:2017].

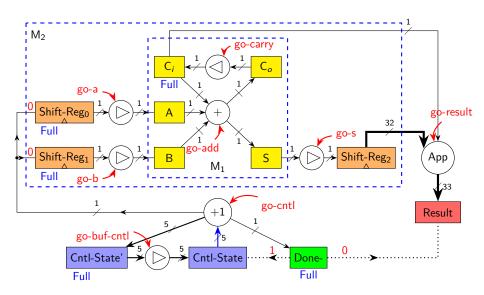
We prove that the self-timed serial adder indeed performs the addition under an appropriate initial condition.

• When the adder finishes its execution, the result is proven to be the sum of the two 32-bit input operands and the carry-in.

Our verification approach applies compositional reasoning.

- Divide the adder's execution into two parts: the loop part and the exit part (the execution after exiting the loop),
- Formalize a loop invariant for the loop part and the adder behavior during the exit part,
- Prove the functional correctness of the adder by glueing these two parts together.

Data Flow of a 32-Bit Self-Timed Serial Adder



Correctness Theorems

Theorem 1 (Partial correctness).

$$async_serial_adder(netlist) \land \qquad (6)$$

$$init_state(st) \land \qquad (7)$$

$$(operand_size = 32) \land \qquad (8)$$

$$interleavings_spec(input_seq, operand_size) \land \qquad (9)$$

$$(st' = run(netlist, input_seq, st, n)) \land$$
(10)

$$full(result_status(st'))$$

$$(11)$$

$$\Rightarrow (\textit{result_value}(\textit{st}') = \textit{shift_reg_0_value}(\textit{st}) + \\ \textit{shift_reg_1_value}(\textit{st}) + \\ \textit{ci_value}(\textit{st}))$$

Correctness Theorems

Theorem 2 (Termination).

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Future Work

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For termination proofs, we need a constraint on **go** signals guaranteeing that **delays are bounded**.

We intend to follow a **hierarchical approach** to prove module-level properties of iterative circuits of the following form:

• Given an initial state of the module, the module's **final state** meets its specification after that module completes execution.

Conclusions

We have presented a framework for modeling and verifying self-timed circuits using the DE system.

Our goal is to develop a methodology that is capable of verifying the functional correctness of self-timed circuit designs at large scale.

 This work also provides libraries for analyzing self-timed systems in ACL2.

We model self-timed systems as networks of links communicating with each other locally via joints, using the link-joint model introduced by Roncken et al.

We also model the **non-determinism of event-ordering** in self-timed circuits by associating each joint with an external go signal.

Our key proof techniques are hierarchical reasoning, compositional reasoning, and induction.

References



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Questions?